FSMD examples in Gezel and in VHDL

Tutorial 06 on Dedicated systems

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tutorial outline

this tutorial deals with:

- hardware implementation methods for FSMD models
- FSMD design and implementation example
  - design of a median computation datapath
  - sequentialization by means of an FSMD
- FSM description in VHDL
  - description styles
  - state encoding
  - delay optimization
- lab experience:
  - design and FPGA implementation of a Collatz delay testbench

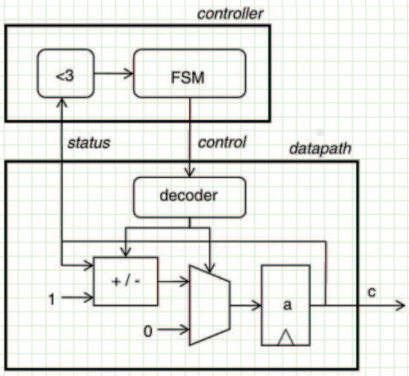
hardware implementation of FSMD models

hardware mapping of datapath expressions: same basic rules as with `always` blocks, however:

hardware datapath structure also depends on instruction schedule by FSM...

why?

consider the up-down counter example from the previous lecture:



Schaumont, Figure 5.8 - Implementation of the up-down counter FSMD

- by the FSM model, different instructions will always execute in different clock cycles
- instructions which are never concurrent may *share* operators' hardware
  such as the adder/subtractor in this case
- a local decoder converts the instruction encoding into control signals for the datapath, to enable the proper execution path
- a local decoder in the controller extracts datapath state information for the FSM conditions

FSMD design example: median computation

function specification: computation of the median in a list of five numbers

problem: design a *stream*-processing filter that produces a new output upon each new input

such filters find application in image processing, for noise reduction

for a fast algorithm, with no list sorting:

in a $(2n+1)$-element list, with pairwise distinct elements, the median has n smaller elements

```
dp median(in a1, a2, a3, a4, a5 : ns(32); out q1 : ns(32)) {
  sig z1, z2, z3, z4, z5, z6, z7, z8, z9, z10 : ns(3);
  sig s1, s2, s3, s4, s5 : ns(1);
  always {
    z1 = (a1 < a2);
    z2 = (a1 < a3);
    z3 = (a1 < a4);
    z4 = (a1 < a5);
    z5 = (a2 < a3);
    z6 = (a2 < a4);
    z7 = (a2 < a5);
    z8 = (a3 < a4);
    z9 = (a3 < a5);
    z10 = (a4 < a5);
    s1 = ((    z1 +     z2 +     z3 +     z4) == 2);
    s2 = (((1-z1) +     z5 +     z6 +     z7) == 2);
    s3 = (((1-z2) + (1-z5) +     z8 +     z9) == 2);
    s4 = (((1-z3) + (1-z6) + (1-z8) +    z10) == 2);
    q1 = s1 ? a1 : s2 ? a2 : s3 ? a3 : s4 ? a4 : a5;
  }
}
```

Schaumont, Listing 5.19 - GEZEL Datapath of a median calculation of five numbers

the algorithm presented by this datapath also works when some elements are equal

- the description is almost identical to that of a C function for the same algorithm
- but with a substantial, practical difference:

  sequential execution of assignments in C vs. parallel execution in the datapath

  where each output production requires just one clock-cycle!

  provided all input values are simultaneously available ...

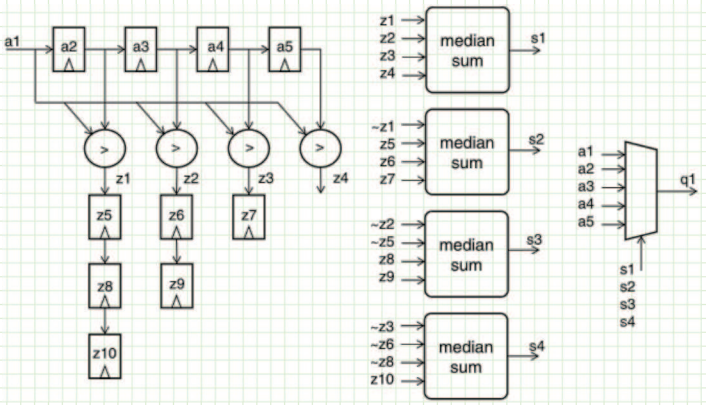
  an FSMD may allow a big space saving, as we are going to see

datapath of a median computation datapath

the presented algorithm requires 192 I/O lines and 10 comparators

a stream processing filter may reduce these requirements to 64 I/O lines and 4 comparators

to this end the hardware stores the four data preceding the last one from the input stream and at every iteration with a new input element reuses the stored results of six comparisons from the previous three iterations



Schaumont, Figure 5.9 - Median-calculation datapath for a stream of values

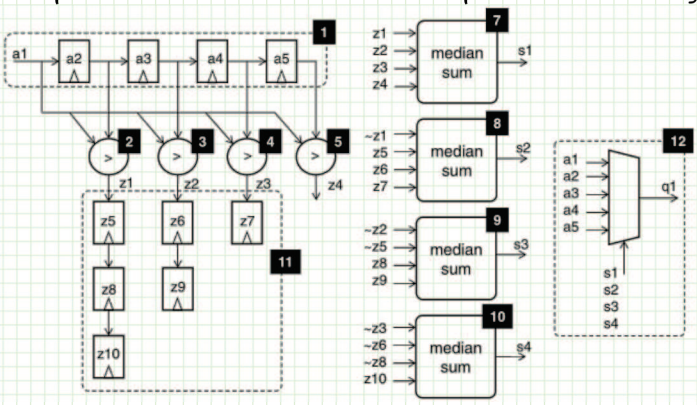
```
dp median(in a1 : ns(32); out q1 : ns(32)) {
  reg a2, a3, a4, a5 : ns(32);
  sig z1, z2, z3, z4;
  reg z5, z6, z7, z8, z9, z10 : ns(3);
  sig s1, s2, s3, s4, s5 : ns(1);
  always {
    a2 = a1;
    a3 = a2;
    a4 = a3;
    a5 = a4;
    z1 = (a1 < a2);
    z2 = (a1 < a3);
    z3 = (a1 < a4);
    z4 = (a1 < a5);
    z5 = z1;
    z6 = z2;
    z7 = z3;
    z8 = z5;
    z9 = z6;
    z10 = z8;
    s1 = ((    z1 +     z2 +     z3 +     z4) == 2);
    s2 = (((1-z1) +     z5 +     z6 +     z7) == 2);
    s3 = (((1-z2) + (1-z5) +     z8 +     z9) == 2);
    s4 = (((1-z3) + (1-z6) + (1-z8) +    z10) == 2);
    q1 = s1 ? a1 : s2 ? a2 : s3 ? a3 : s4 ? a4 : a5;
  }
}
```

sequentialization by means of an FSMD

the filter in fig 5.9 accepts a new input and produces a new output at every clock cycle

the number of computing components may be further reduced by distributing the computation over multiple clock cycles under a sequential schedule, so that a single comparator and a single module to compute $s1$, $s2$, $s3$, $s4$ are reused, at the cost of adding a few multiplexers and registers for the internal signals

the figure shows the schedule, the FSMD which implements this idea is in Schaumont Sect. 5.5.4, which also presents the testbench FSMD here reproduced aside the figure

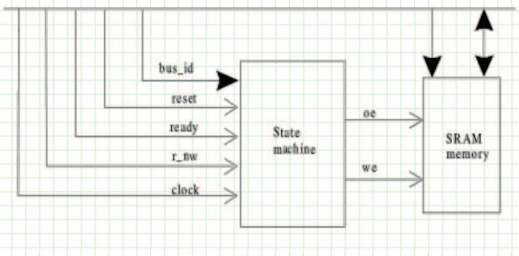


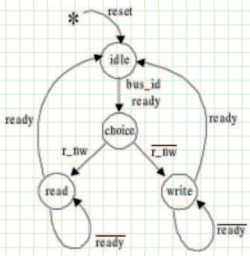
Schaumont, Figure 5.10 - Sequential schedule median-calculation datapath for a stream of values

```
dp t_median {
    sig istr, ostr : ns(1);
    sig a1_in, q1 : ns(32);
    use median(istr, a1_in, ostr, q1);
    reg r : ns(1);
    reg c : ns(16);
    always { r = ostr; }
    sfg init { c = 0x1234; }
    sfg sendin { a1_in = c;
                 c = (c[0] ^ c[2] ^ c[3] ^ c[5]) # c[15:1];
                 istr = 1; }
    sfg noin { a1_in = 0;
               istr = 0; }
}
fsm ctl_t_median(t_median) {
    initial s0;
    state s1, s2;
    @s0 (init, noin) -> s1;
    @s1 (sendin) -> s2;
    @s2 if (r) then (noin) -> s1;
                else (noin) -> s2;
}
```


FSM description in VHDL

a simple example: memory controller

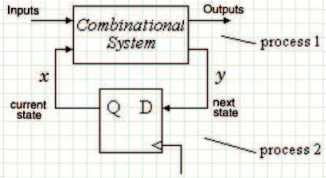




state	outputs oe	we
idle	0	0
choice	0	0
read	1	0
write	0	1

State transition diagram of a memory controller

general FSM hardware structure



FSM hardware structure

in the VHDL description, states form an enumerated type (see enclosed code):

- *two-process*: most immediate way to reflect the scheme in the figure, one process is sensitive to the clock
  - this is the translation style with `fdlvhd`
- *one-process*: just one state signal, Moore FSM outputs may be described by concurrent assignments, outside the process
- *three-process*: state transition and output functions described by distinct combinational processes

state encoding techniques

a binary encoding of the state enumerated type is needed to synthesize an HDL description of an FSM automatically generated by the synthesis tool if not explicitly specified in the description

frequently used encodings:

- *sequential*: compact, a set of n states is encoded in the set of binary representations of the natural numbers $< n$, with $\lceil \log_2 n \rceil$ bits
- *one-hot*: with as many bits as are the states, where only one bit is set to 1 in the encoding (bijection between states and bit positions in the encoding), automated synthesis often generates this encoding
- *ad-hoc* to optimize the delay, see next page

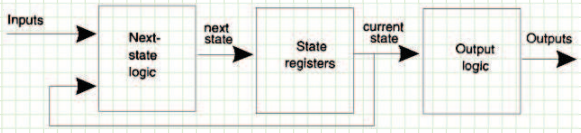
explicit state encoding may be described in VHDL by defining the state type as a subtype of STD_LOGIC_VECTOR and by declaring the state encodings as constants of that type; for example, here are sequential and *one-hot* encodings of the states from the seen example:

```
subtype state is std_logic_vector(1 downto 0);
   constant idle : state := "00";
   constant choice : state := "01";
   constant read : state := "10";
   constant write : state := "11";
```

```
subtype state is std_logic_vector(3 downto 0);
   constant idle : state := "0001";
   constant choice : state := "0010";
   constant read : state := "0100";
   constant write : state := "1000";
```

if the encoding is not surjective, *e.g.* a *one-hot* encoding, the construct `case stato is when ...` must have the final clause `when others ...`

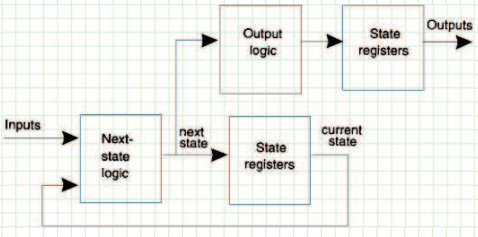
delay optimization



Current-state dependent outputs (Moore FSM)

the block diagram corresponds to the two-process and one-process descriptions in the enclosed code

a significant clock-to-output delay may result

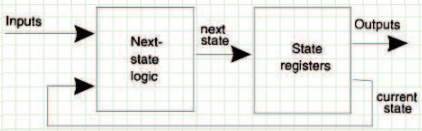


Next-state dependent outputs

this may be reduced to the flip-flop clock-to-q delay by computing outputs based on the next state and by placing registers before outputs

the three-process architecture in the enclosed code exemplifies this optimization

N.B. since processes are *sequential*, this optimization is also applicable to two-process and one-process architectures as well, by deferring outputs' updates until after state updates



Reduction of clock-output delay

another optimization technique consists in encoding states so as to let outputs coincide with some of the (encoded) state bits

e.g. in the seen example:

idle	0 0 0
choice	0 0 1
read	0 1 0
write	1 0 0

lab experience

this experience has a twofold target

1. extension of the dataflow model out of the third lab experience for the delay computation of Collatz trajectories to an FSMD model that would include a pseudorandom generator feeding the datapath with initial values of trajectories, similar to the `sendin` instruction in the testbench example for the median computation filter, and that would alternate the output of the generated initial value with the output of the computed delay for the corresponding trajectory
   for the first target, translation to VHDL and simulation are required, the latter as a correctness check
2. implementation of the model on the DE1-SoC FPGA, with user control inputs to enable the alternating data output on a series of five 7-segment displays, in decimal representation

the I/O for the first target model consist of:

- binary input `start`, to enable the start of a new trajectory, with output of the generated (16-bit) initial value
- binary output `done`, to signal te end of computation of a trajectory delay
- binary input `dout`, to enable the output of the computed delay, on the same port as of the initial value
- `display` output of 16-bit numbers, multiplexed over the alternating output of initial value and delay as mentioned above

after the initial state where the generator is initialized, it is suggested to let the control FSM go through a sequence of four states:

1. wait for user control input `start`
2. output of the trajectory initial value, start up the delay computation, and wait for its completion
3. `done` output and wait for user control input `dout`
4. output of the delay value and back to state 1

for the second target, use can be made of the 7-segment display decoder from the two latest lab experiences, but this is to be fed with BCD representations of the 16-bit output decimal digits: five displays are needed to this purpose and package `bcd`, is available, that provides the `BCD_encode` function
use keys `KEY0`, `KE1`, `KEY2` for inputs `RST`, `start`, `dout`, LED `LEDR0` for output `done`, and `CLOCK_50` for the clock

references

recommended readings:

Schaumont, Ch. 5, Sect. 5.4.4-5.5

readings for further consultation:

C. Brandolese, Introduzione al linguaggio VHDL (Politecnico di Milano), Cap. 8

Zwoliński, Ch. 7, Sect. 7.1-2

useful materials for the proposed lab experience

Note sul VHDL, Corso di Architettura dei Sistemi Integrati, Capp. 10.1, 11, 13

VHDL sources: 7-segment display decoder, package `bcd`

sources of examples in this presentation:

Gezel: parallel Median, sequential Median

VHDL: SRAM controller